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**IN THE SPECIFICATION:****Page 4, lines 11-13**

~~FIG. 1~~ is FIGS. 1A and 1B are a high level block diagrams of a Doppler-corrected rake finger structure used in a communications receiver of the present invention.

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~~FIG. 3~~ is AFIGS. 3A, 3B, and 3C are detailed block diagrams of the Doppler-corrected rake finger structure shown in ~~FIG. 1~~FIGS. 1A and 1B.

**Page 5, lines 24-30 and Page 6, lines 1-2**

As shown in ~~FIG. 1~~FIGS. 1A and 1B, a high level block diagram of a rake receiver 10 having a Doppler-corrected rake finger structure is illustrated. The signal is down converted 10a and descrambled 10b. The signal is next split at baseband via a mixer and phase shift circuit 11 into in-phase (I) and quadrature (Q) components and into in-phase (I) first and second paths and quadrature (Q) first and second paths. The first path includes a pilot channel rake section 12 having I and Q Doppler estimation channels 14, 16 for estimating the Doppler change in frequency based on a common pilot channel.

**Page 7, lines 25-30**

For purposes of explanation, a description of the algorithm used with differential detection in the rake receiver design shown in FIGS. 1 and 3 is set forth in detail, followed by a detailed description of the circuit shown in ~~FIG. 3~~FIGS. 3A through 3C that implements the method and algorithm.

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Let  $r_I(t)$  and  $r_Q(t)$  represent the in-phase and quadrature part of the received signal for another rake finger that is used to receive the data channel, such as shown in ~~FIGS. 1 and 3~~FIGS. 1A, 1B, 3A, 3B, and 3C. Therefore:

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$$r_1(t) = \sqrt{2p\alpha(t)a_1(t)c_1(t)} \sin(2\pi\Delta f t + \theta(t)) + n_1(t)$$

$$r_Q(t) = \sqrt{2p\alpha(t)a_Q(t)c_{Q1}(t)} \cos(2\pi\Delta f t + \theta(t)) + n_Q(t)$$

Equation 6

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~~FIG. 3~~FIGS. 3A through 3C ~~show~~show a more detailed block circuit diagram of the rake finger structure shown in FIG. 1. There is no illustrated acquisition and tracking circuit as would typically be used in a rake receiver. The phase error introduced due to the imperfect acquisition and tracking is considered in  $\theta(t)$ , which can be alleviated via simple averaged channel estimation. There is a  $2NT_c$  delay in the other fingers, because it takes  $2NT_c$  samples to obtain  $I_k$ ,  $I_{k+1}$ ,  $Q_k$ , and  $Q_{k+1}$ .